

# Single Photon Counting in the Visible



L. Strüder, M. Porro, G. De Vita, S. Herrmann,  
J. Treis, S. Wölfel  
Max-Planck-Institut f. extraterrestr. Physik  
Garching



R. H. Richter  
Max-Planck-Institut f. Physik  
Munich



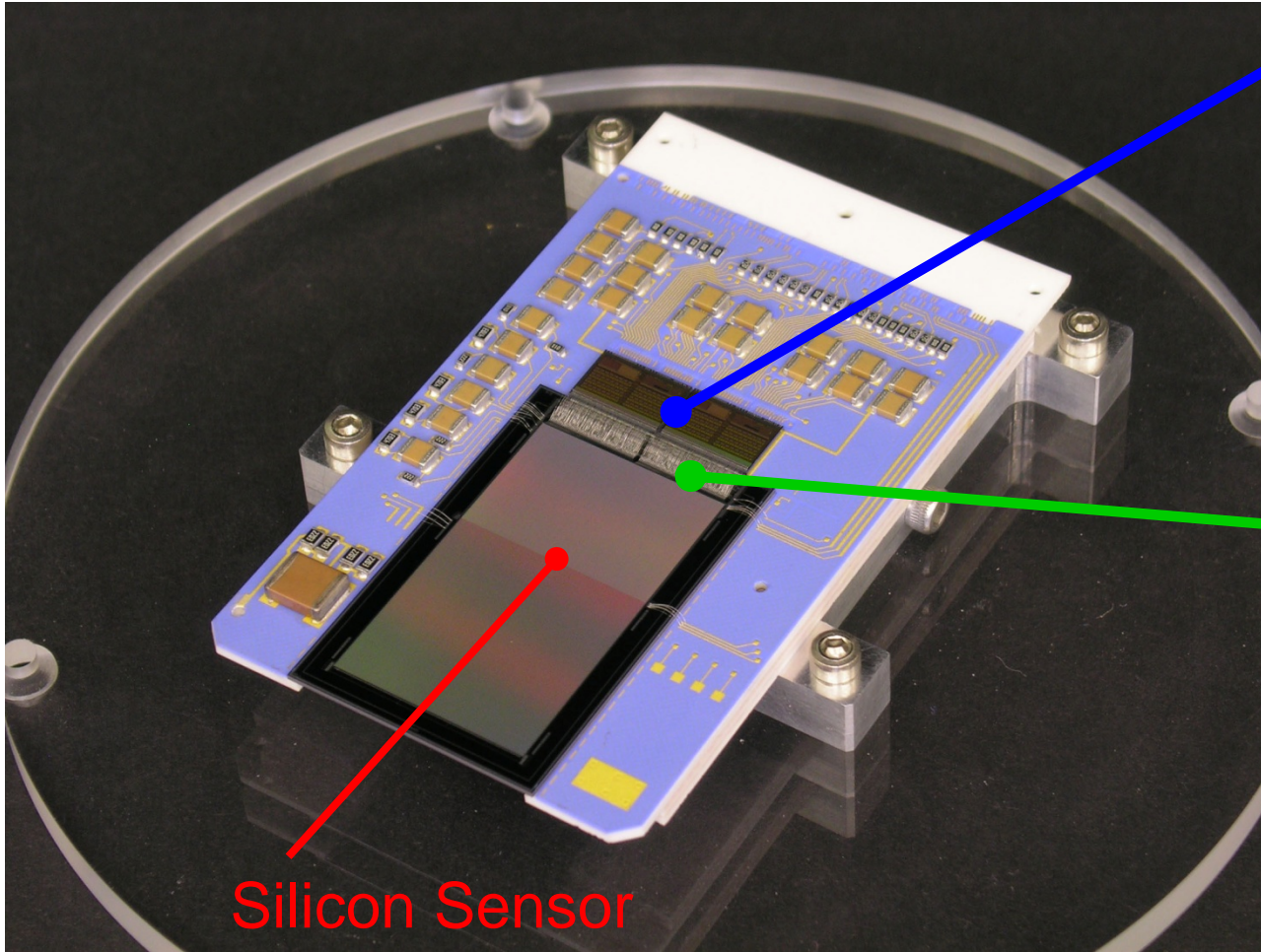
P. Lechner, G. Lutz,  
PNSensor GmbH  
Munich

**PNSensor**

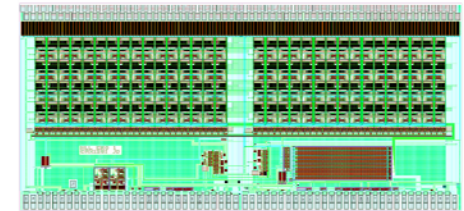
## OUTLINE

- System Definition
- DePMOS and RNDR Device Concept
- RNDR working principle
- Experimental results
- Gatable APS devices
- Achieved and achievable performance
- Conclusions

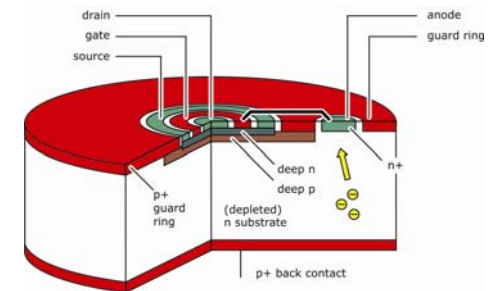
# System Definition



Multi-Channel readout ASIC performing time variant filtering



First amplification device array integrated on the sensor

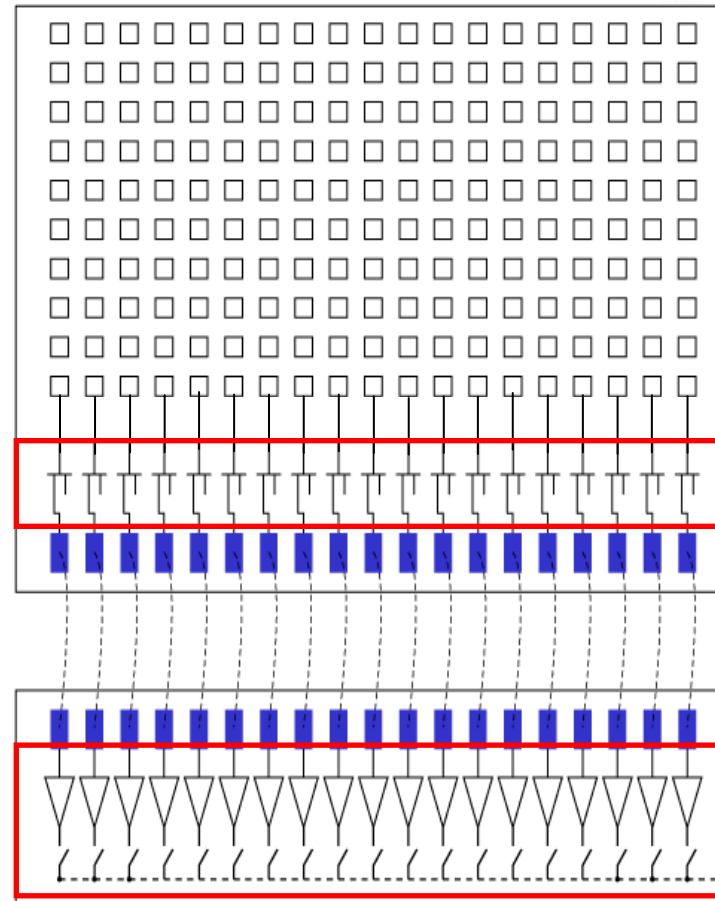


**Silicon Sensor**  
(image and frame store area)

# pnCCD for single photon counting



- High readout speed (1 kHz at 256x256 formats)
- High quantum efficiency
- High charge transfer efficiency
- High spectroscopic resolution with X-rays (2 electrons r.m.s. @ - 60 °C, TEC)
- Integrated JFET as first amplification stage on every readout anode
- Parallel Readout (one complete readout channel per column)
- With anti-reflecting coating it is possible to achieve a quantum efficiency close to 100% for near infrared (300-1100 nm)
- 2 el. r.m.s. would not allow to detect less than 10 optical photons



*Image +  
frame store  
area*

*(256x256  
+  
256x256)*

*Integrated  
SSJFETs*

*Bond wires*

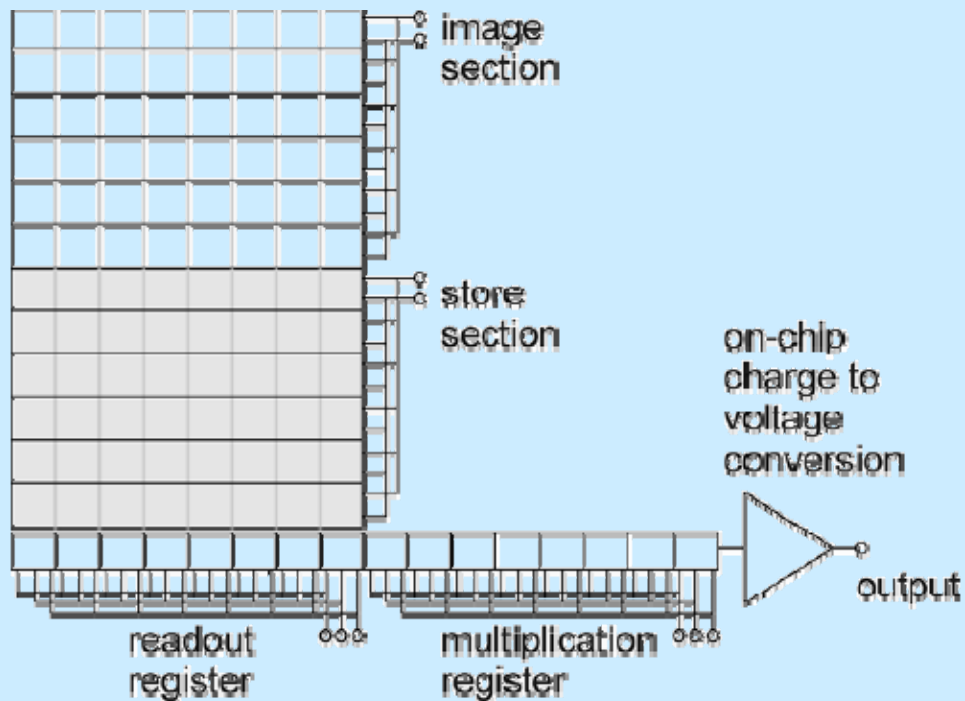
*Time variant  
Readout ASIC*

To achieve single photon resolution:

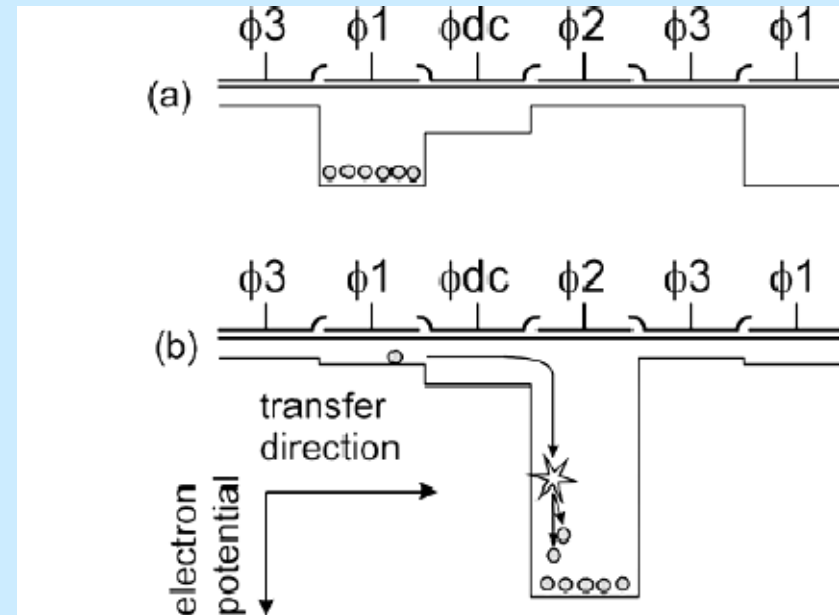
- use avalanche multiplication
- the on-Detector JFETs can be replaced by RNDR-DePMOS
- a suitable Multi-channel readout ASIC must be implemented

# EMCCDs principles

## EMCCD

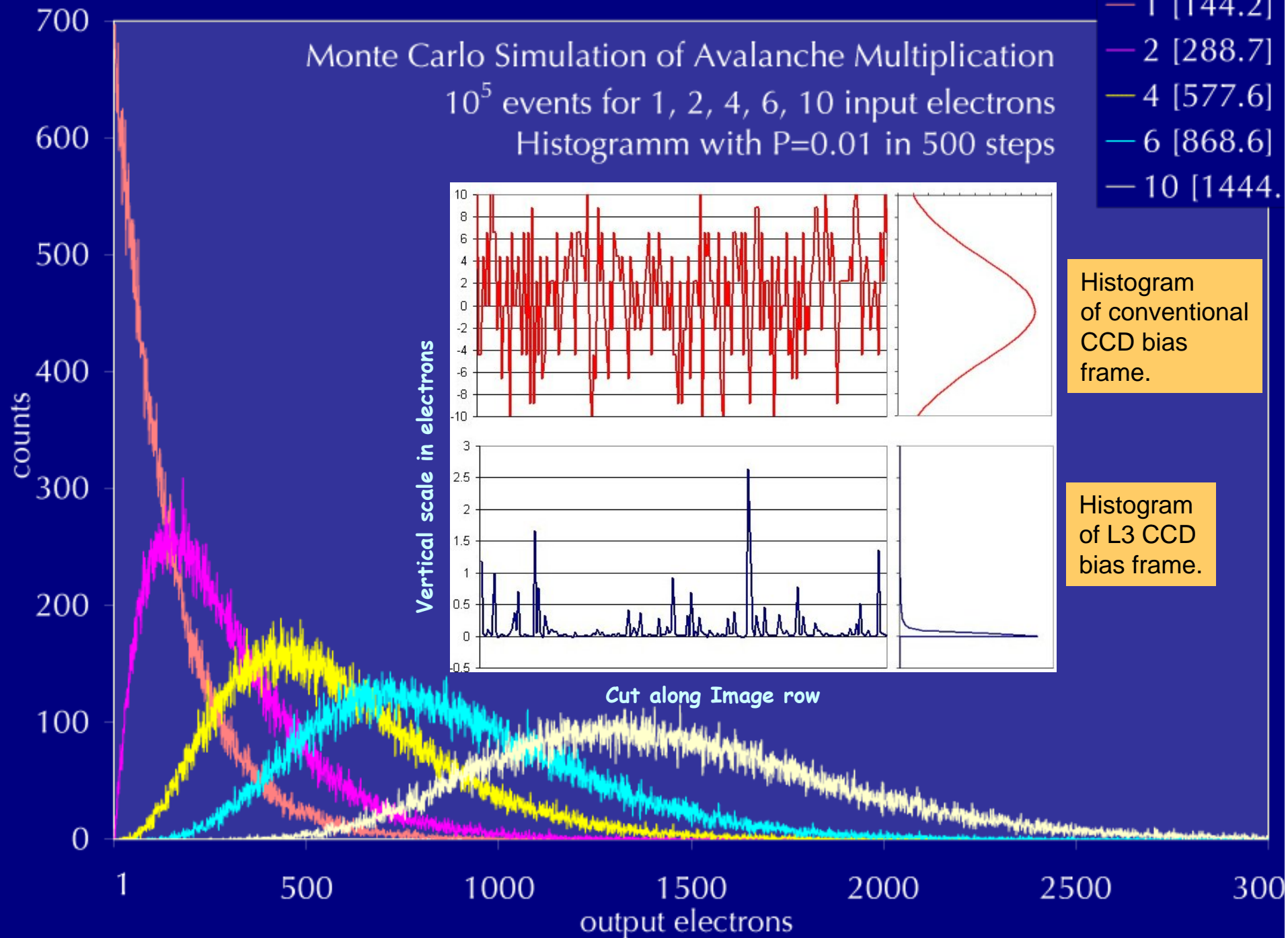


## gate structure

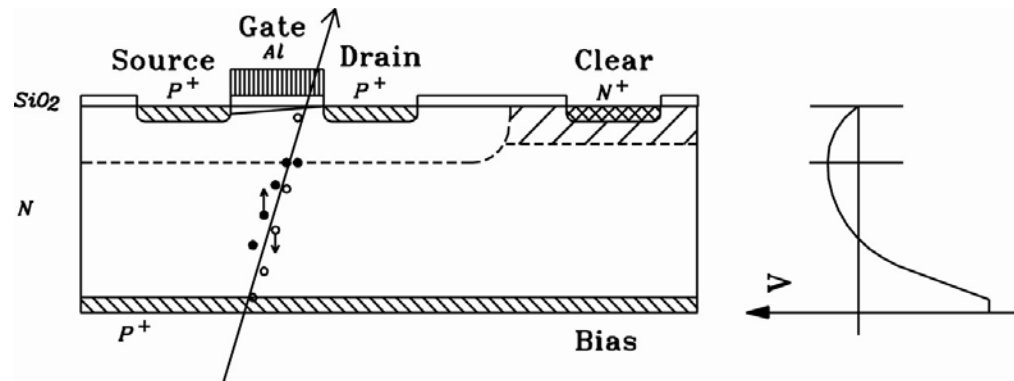


Monte Carlo Simulation of Avalanche Multiplication  
 $10^5$  events for 1, 2, 4, 6, 10 input electrons  
Histogramm with  $P=0.01$  in 500 steps

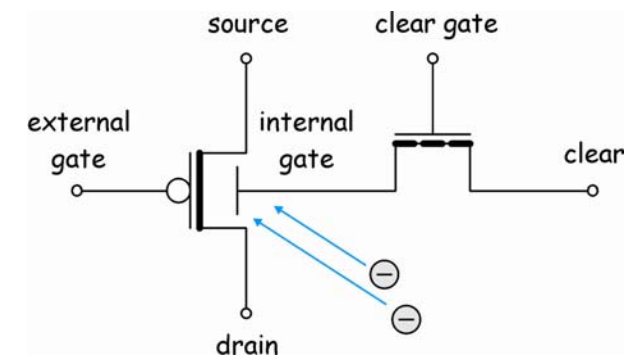
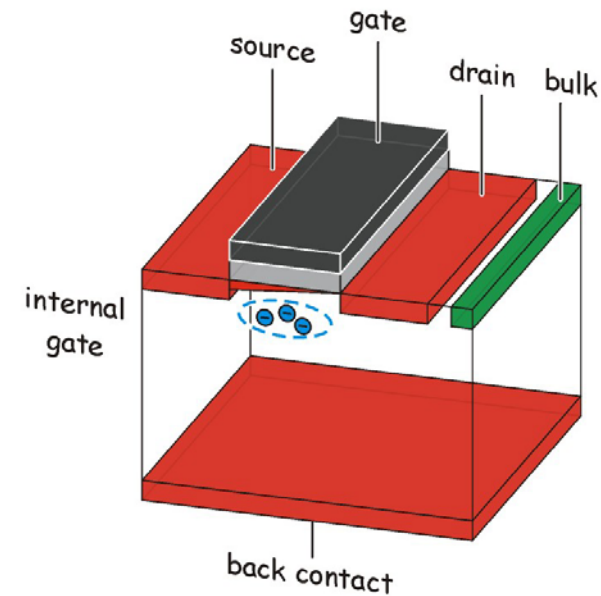
- 1 [144.2]
- 2 [288.7]
- 4 [577.6]
- 6 [868.6]
- 10 [1444.2]



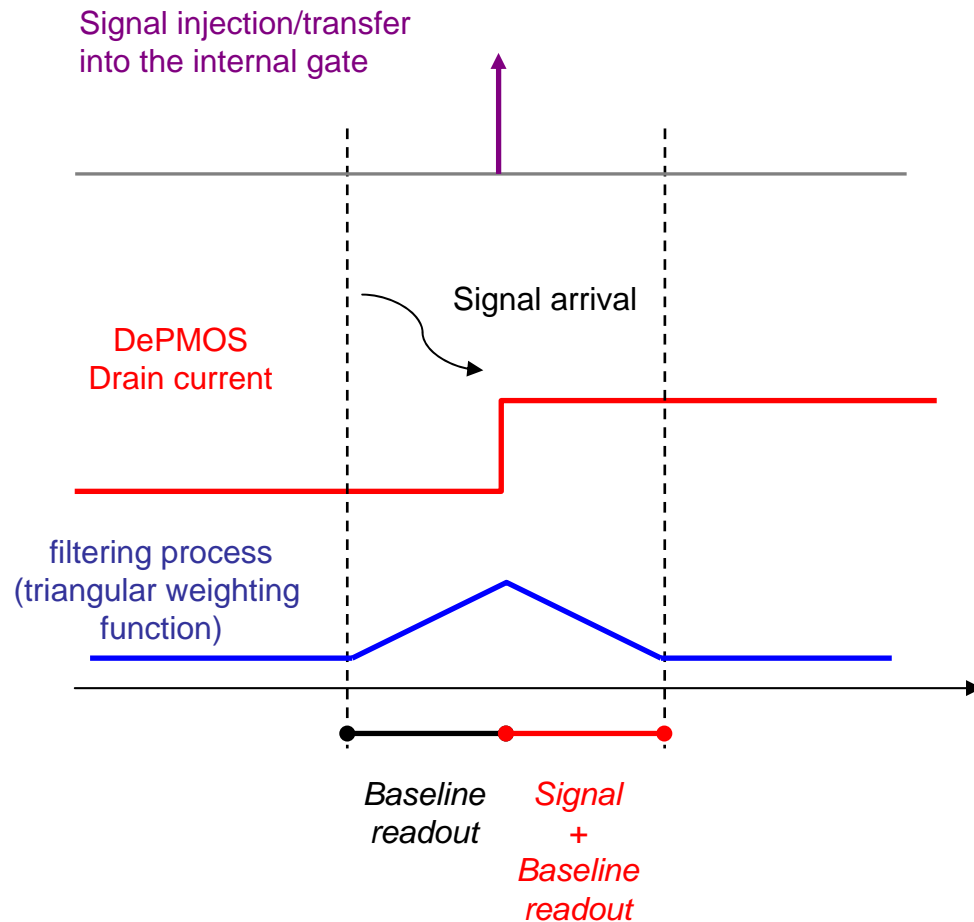
# DePMOS concept



- p-channel MOSFET integrated on high-ohmic, sideward depleted n-substrate
- a potential minimum is formed by S/D potentials aided by a deep n implantation
- electrons are collected in an internal gate close to the surface
- the transistor current is modulated by charge collected in the internal gate
- the transistor can be switched on/off by an external (top) gate
- An n+ clear contact surrounded by a clear gate is used to remove the charge from the internal gate

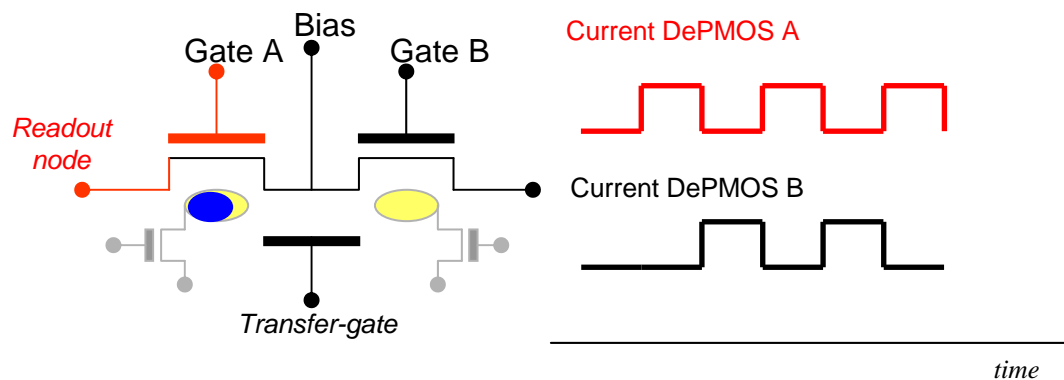
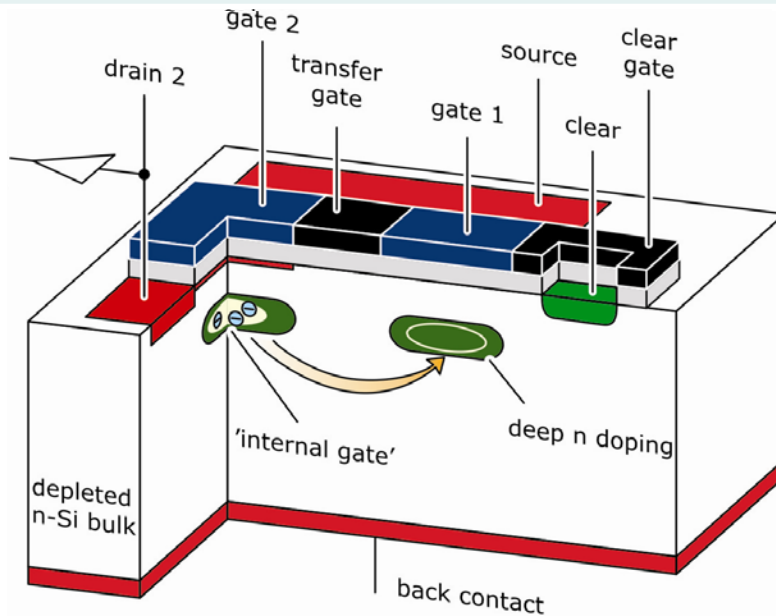


# DePMOS Readout



- The signal arrival time is known  
(the charge is transferred from the CCD column and then switched between the two pixels clocking the transfer gate)
- One measurement is composed of the difference of two evaluations:
  - Baseline
  - Baseline + signal
- A time variant filter is used
- The triangular weighting function is the time limited optimum filter for voltage noise

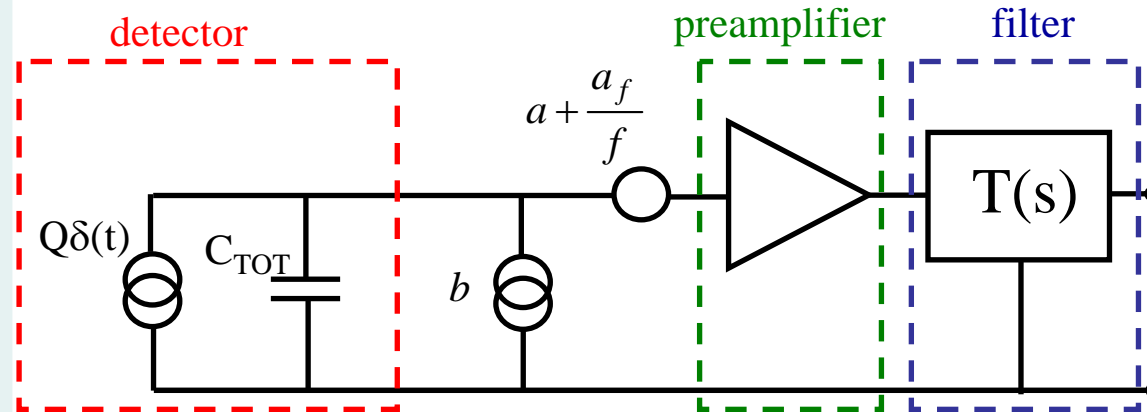
# RNDR DePMOS



- RNDR Device is composed of 2 adjacent DePMOS structures
- The charge in the internal gate can be switched between the internal gates of the two DEPMOS, thanks to one (or more) *transfer gate(s)*
- When the internal gate of one device is full the internal gate of the other one is empty
- Moving the signal charge from one device to the other allows to reproduce the signal arbitrary often
- The main limitation is given by the leakage current that fills the internal gate
- It is possible to read out the signal from both devices or to use one device just as a storage for the charge



# ENC evaluation



white voltage noise component decreases as measurement time increases

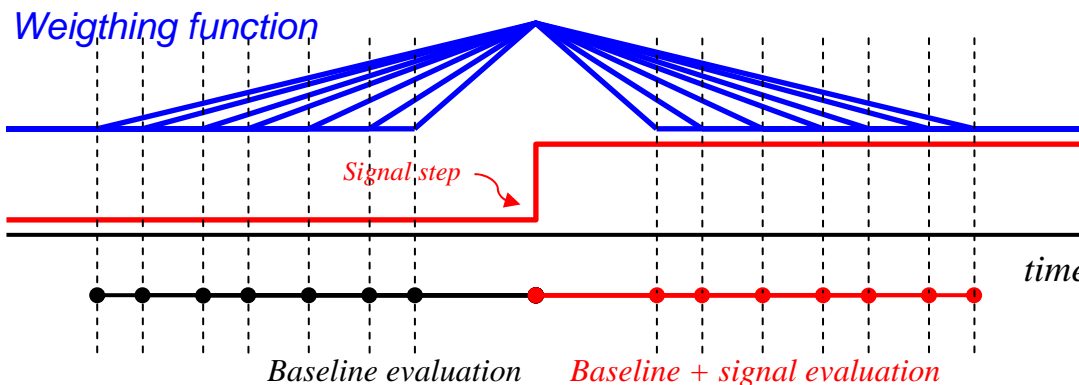
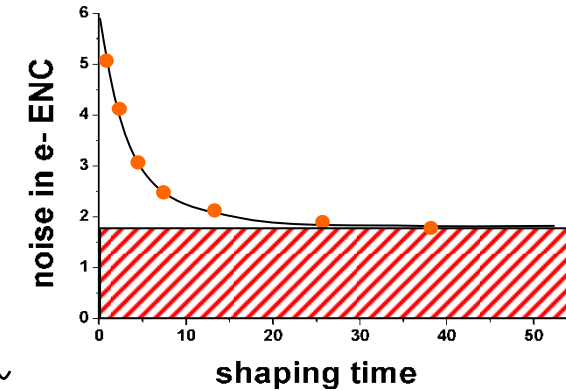
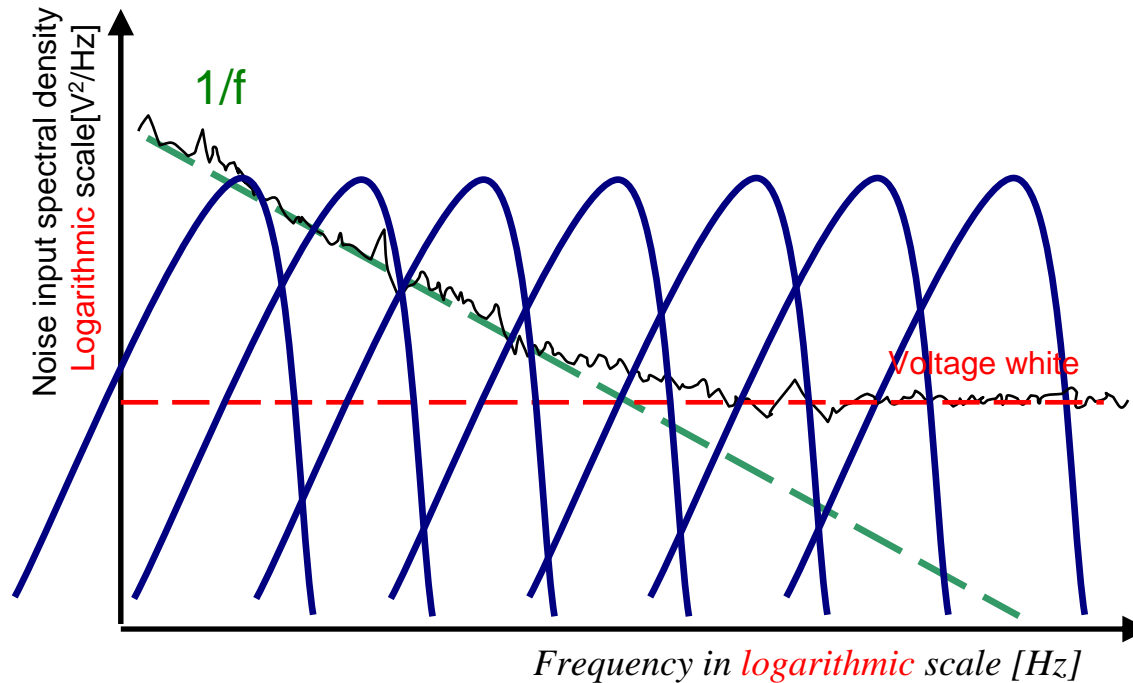
1/f voltage noise component is independent from measurement time

Current white noise (lorentian noise) increases with measurement time

$$ENC^2 = \frac{a}{\tau} C_{TOT}^2 A_1 + 2\pi a_f C_{TOT}^2 A_2 + b \tau A_3$$

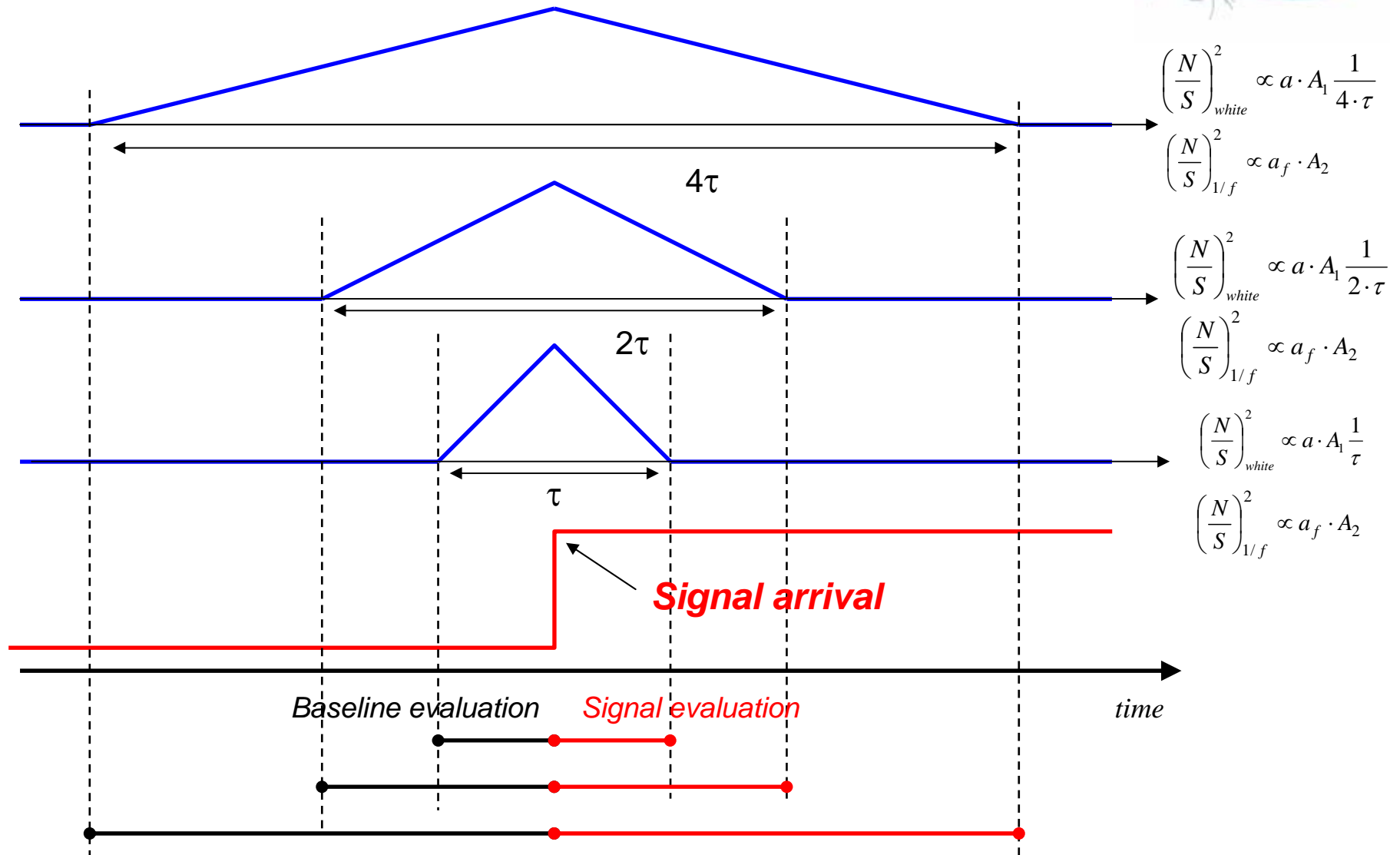
- $C_{TOT}$  DePMOS equivalent input capacitance
- $a, a_f, b$  physical noise sources
- $A_1, A_2, A_3$  filter parameters
- $\tau$  filter shaping time

# Single Readout noise filtering



- The weighing function is fixed
- The measurement time changes (shaping time)
- This means to translate the transfer function in the frequency plot in LOGARITHMIC scale
- The same signal amplitude is measured for different WFs
- Different WFs integrate different regions of the input noise spectral density
- Every decade of the spectrum contains the same r.m.s. value of 1/f noise

# Single Readout noise filtering

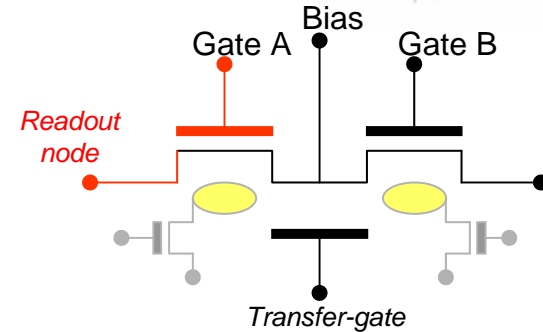
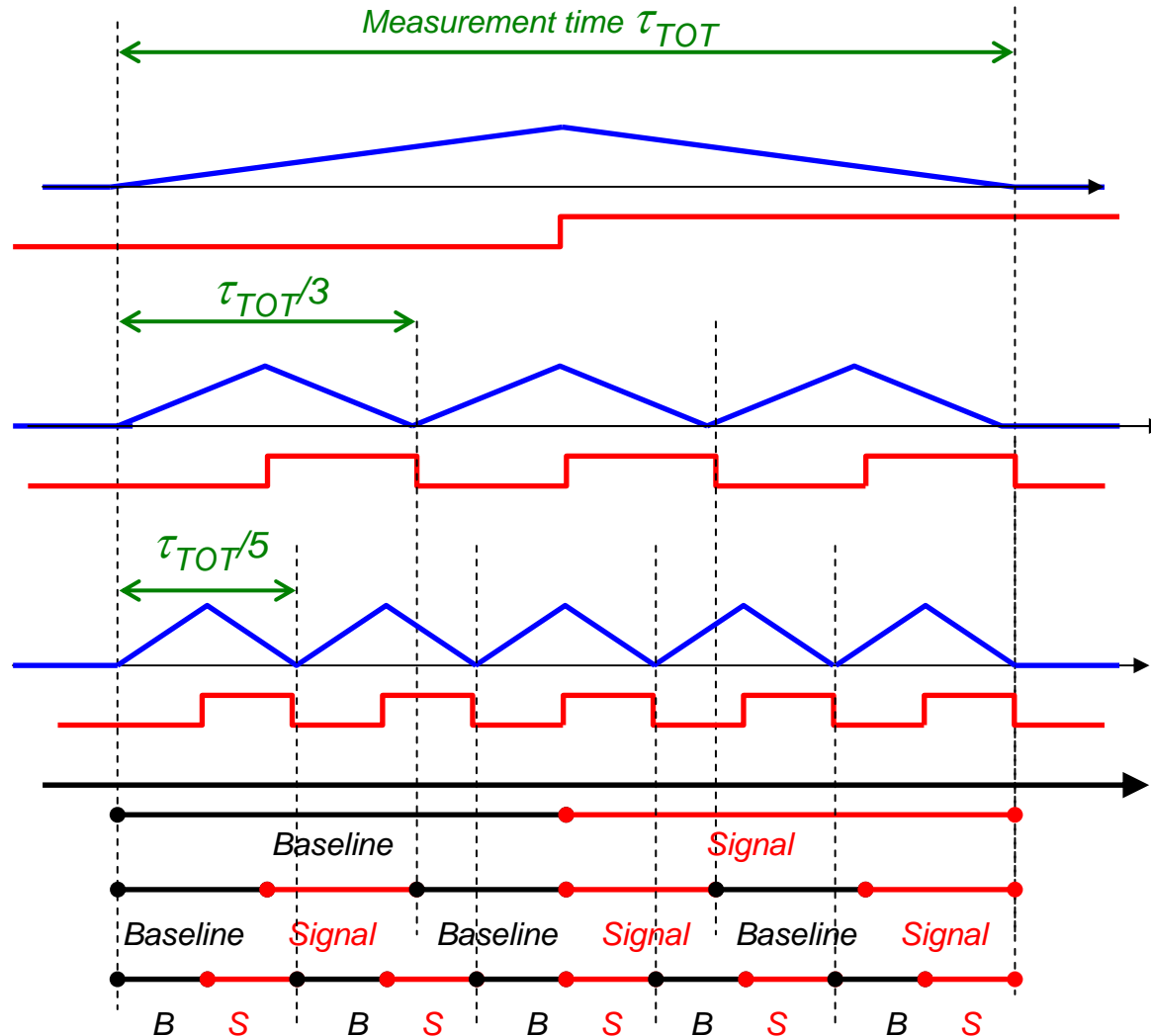


# Repetitive non Destructive Readout



- We fix the total measurement time  $\tau_{TOT}$  (e.g. such that the  $1/f$  noise is dominant)
- We reproduce the signal  $n$  times, moving the charge back and forth from the internal gate of one DePMOS to the internal gate of the other one.
- We measure the signal  $n$  times and we make an average of the measurements
- The signal we reproduce is always the same, i.e. the signal charge is not spoiled by leakage current electrons that can cumulate in the internal gate
- Every signal measurement is the *diffence* of the “baseline” and the “baseline+signal” evaluation (that is why we need to move the signal charge back and forth from one DePMOS to the other one)
- Since  $\tau_{TOT}$  fixed the time available for each single measurement is  $\tau_{TOT}/n$
- The noise of the  $n$  measurements sums up quadratically
- The signal sums up linearly

# Repetitive Non Destructive Readout



- We readout one DePMOS, e.g. DePMOS A
- The other DePMOS is used to store the signal charge when the baseline of DePMOS A is evaluated
- The signal can be reproduced (transferred back and forth into DePMOS A)  $n$  times

$$\left(\frac{N}{S}\right)_{\text{White}}^2 \propto \frac{S_{A_1} \left( \frac{A_1^1 3}{\tau_{TOT} \tau_{TOT}} \right)}{A_1 (\text{signal})^2} = \frac{1}{A_1} \frac{1}{\tau_{TOT} \tau_{TOT}}$$

$$\left(\frac{N}{S}\right)_{1/f}^2 \propto \frac{\hbar \cdot A_2}{(\text{signal})^2 \hbar} = A_2$$

# RNDR: properties and considerations



When total measurement time is fixed:

- White noise is independent from the number of measurements
- $1/f$  scales approximately as  $1/n$   
(It scales as  $1/n^x$  where  $x$  is close to 1. For an exact calculation see: E. Gatti et. Al. "Multiple read-out of signals in presence of arbitrary noises. Optimum filters", NIM A 417, 1998)
- White current noise scales like  $1/n^2$ .
- The noise relative to the leakage current in the internal gate does not scale and increases with the total measurement time (see S. Woelfel et al. "A Novel Way of Single Optical Photon Detection: Beating the  $1/f$  Noise Limit With Ultra High Resolution DEPFET-RNDR Devices", IEEE TNS Volume 54, issue 4, Part 3, Aug. 2007 )

When the single measurement time is fixed (total time increases with number of measurements):

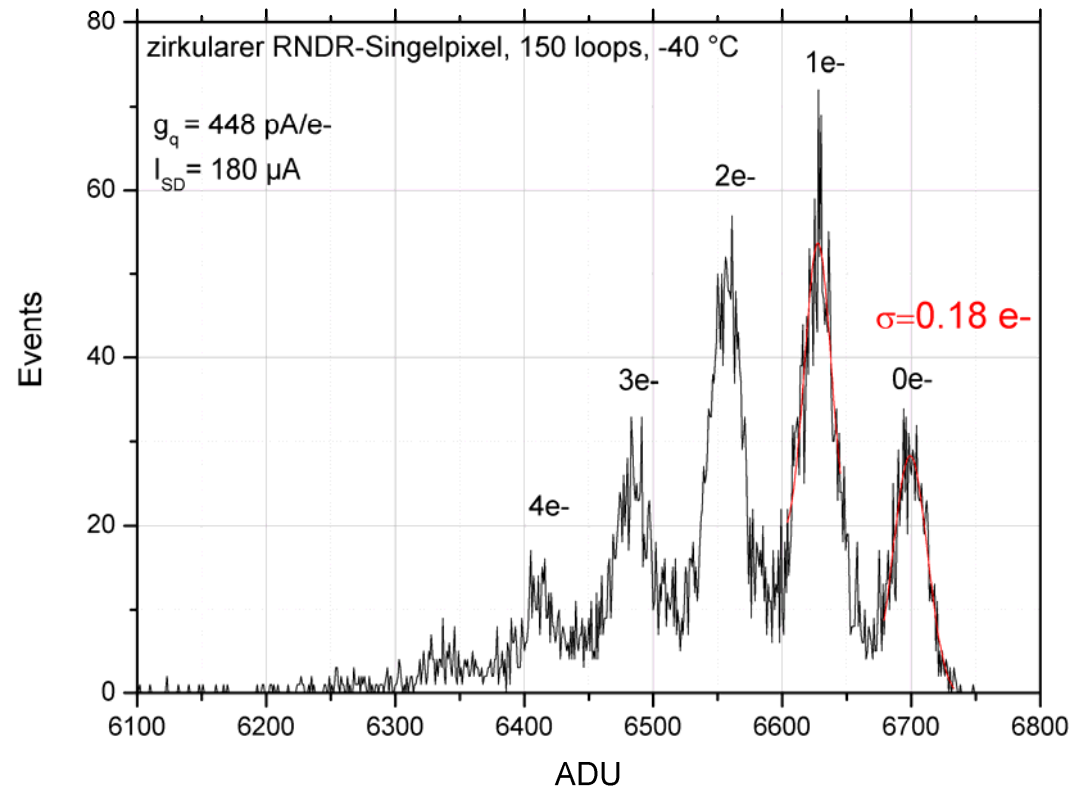
- all the three components scale as  $1/n$
- the noise of the leakage current in the internal gate increases with the measurement time

- RNDR must be used when  $1/f$  noise is dominant
- It is possible :
  - 1) to increase the total measurement time to make white voltage noise negligible
  - 2) to use multiple readout to decrease  $1/f$  noise contribution
- The total measurement time is limited by:
  - experimental constraints
  - leakage current that fills the internal gate
- The system properties are tunable: it is possible to trade speed with resolution

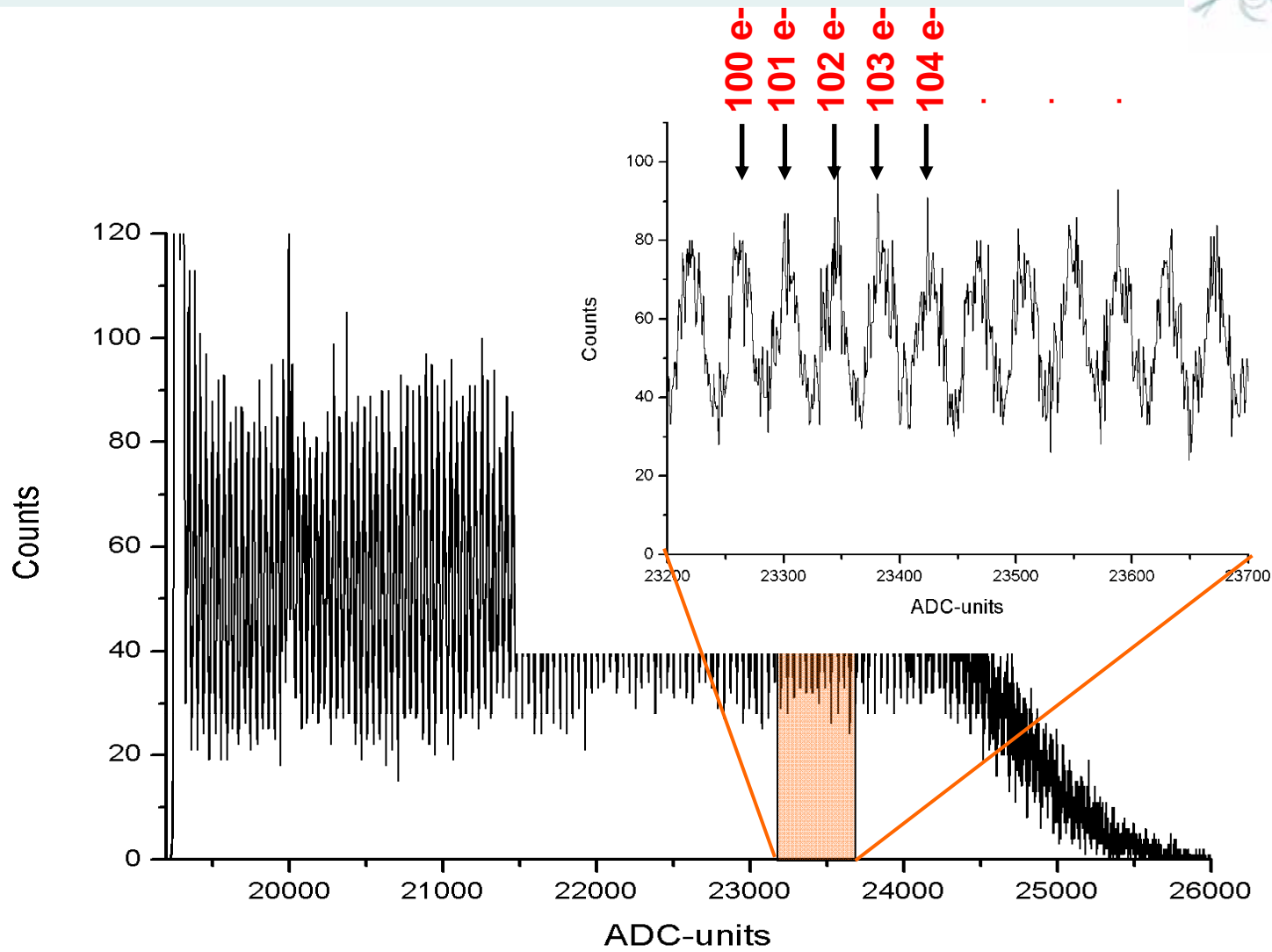
# RNDR-DEPFET



- Single pixel structure: ENC for one readout: 2.1 electrons (rms)
- Smaller gate-length  $\rightarrow$  higher  $g_q$   $\rightarrow$  higher S/N

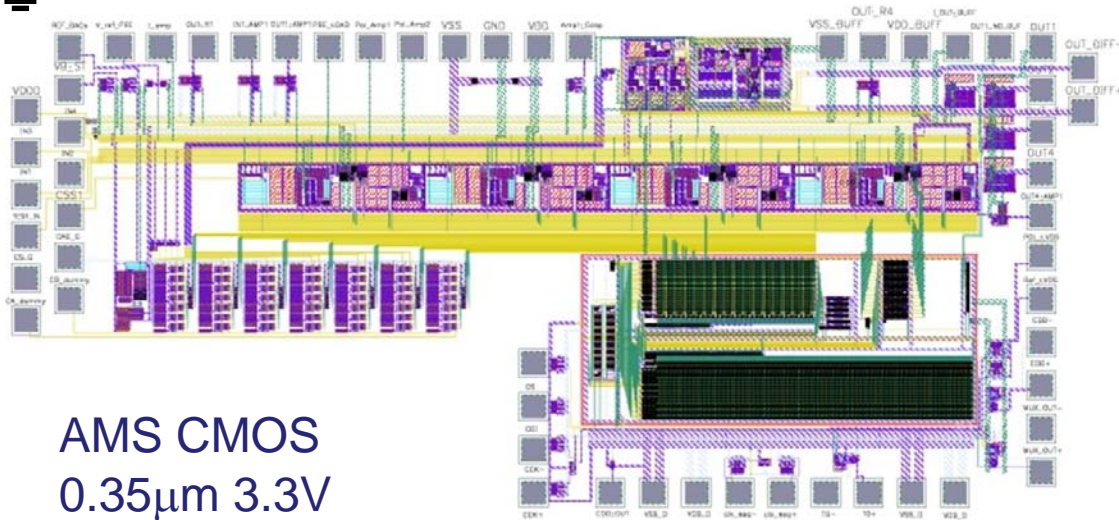
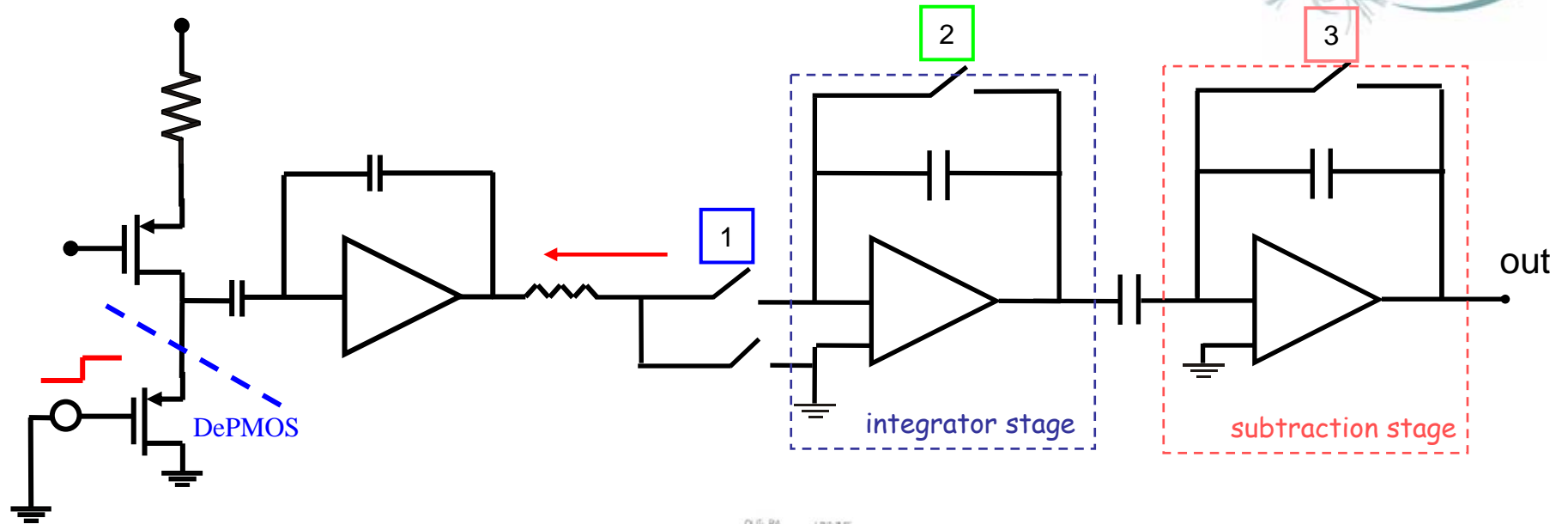


# How to distinguish 100 electrons from 101?

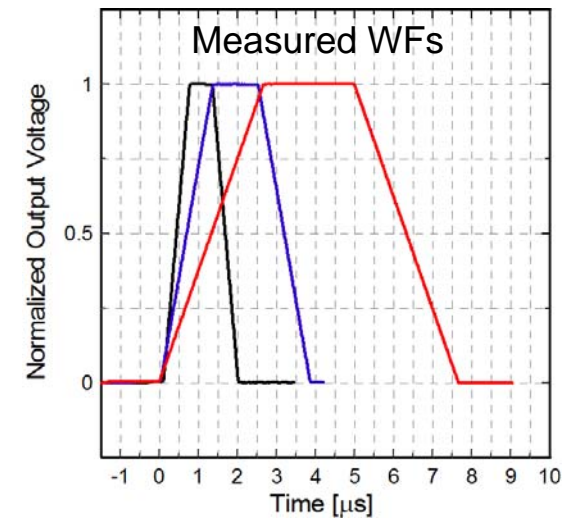




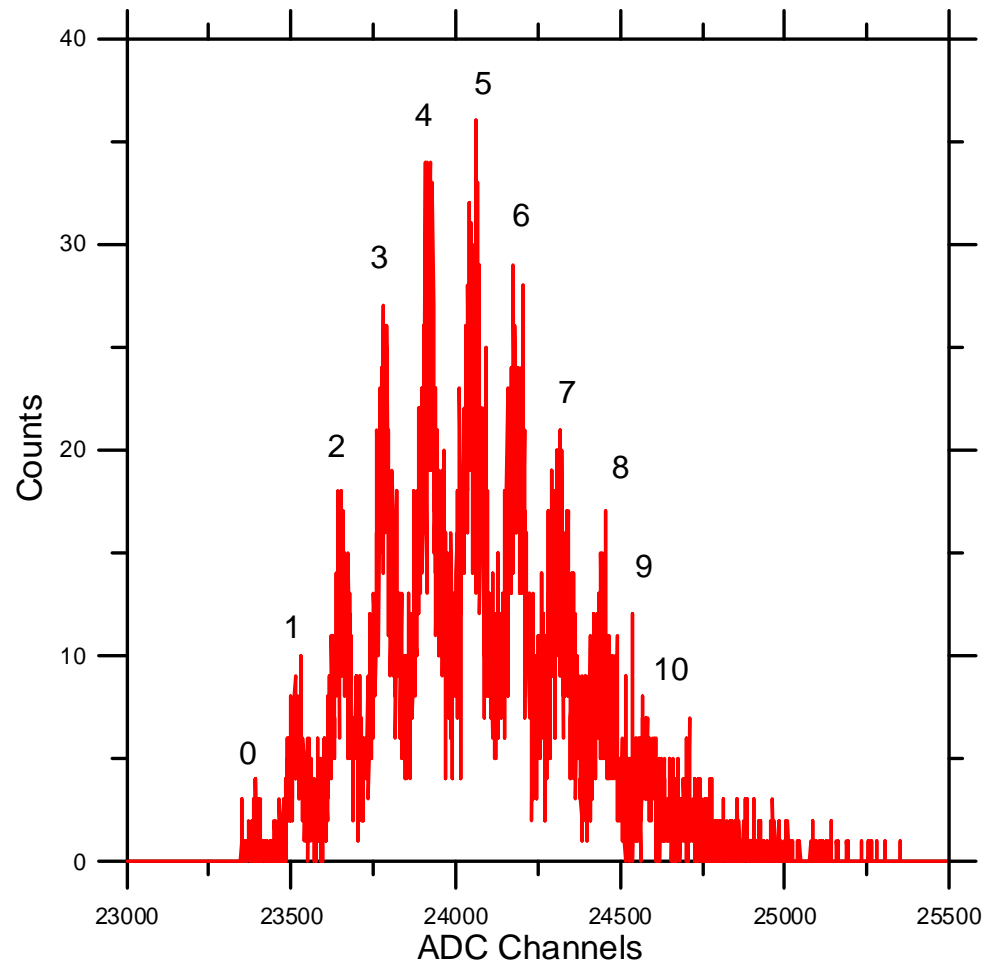
# Trapezoidal Filter Readout ASIC



AMS CMOS  
0.35µm 3.3V

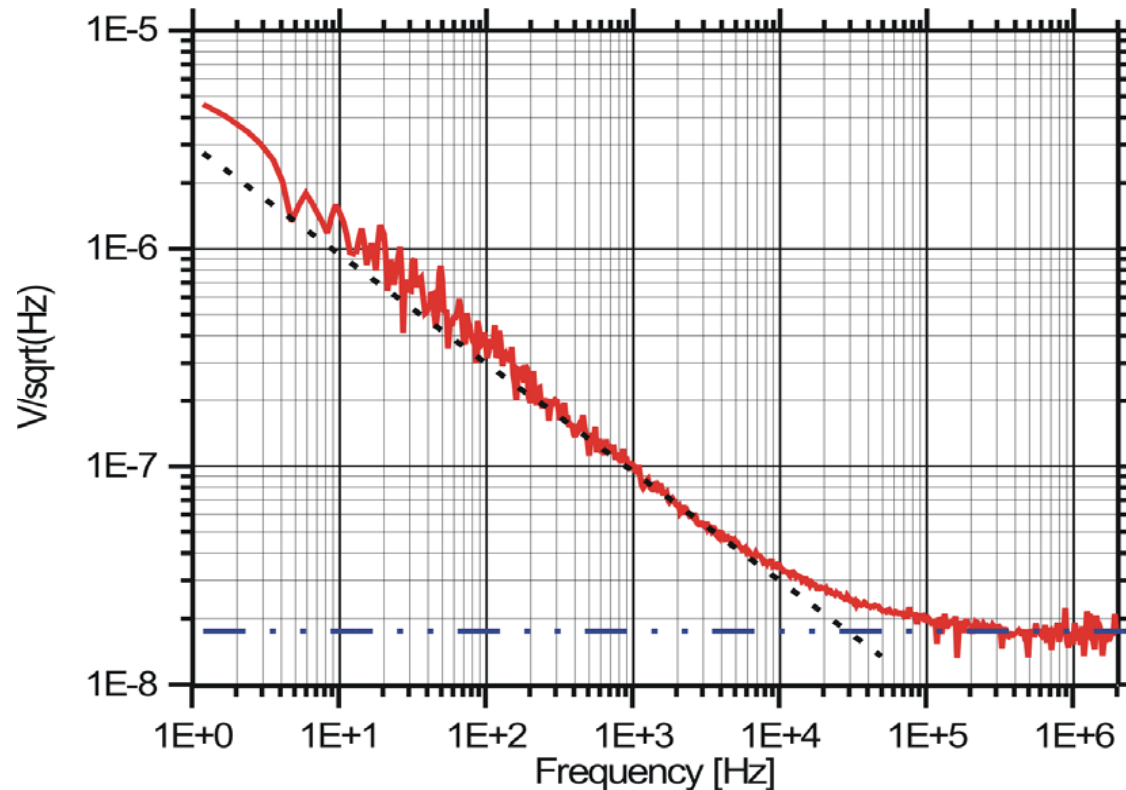


# Single Photon Resolution



- A weak intensity laser has been used to inject electrons into the RNDR-DePMOS
- The laser injects in average 5 electrons
- $T = -50\text{ }^{\circ}\text{C}$
- A trapezoidal weighting function has been used with a total processing time of  $20\text{ }\mu\text{s}$
- ENC  $\sim 0.25\text{ el. r.m.s.}$

# Achievable Performance



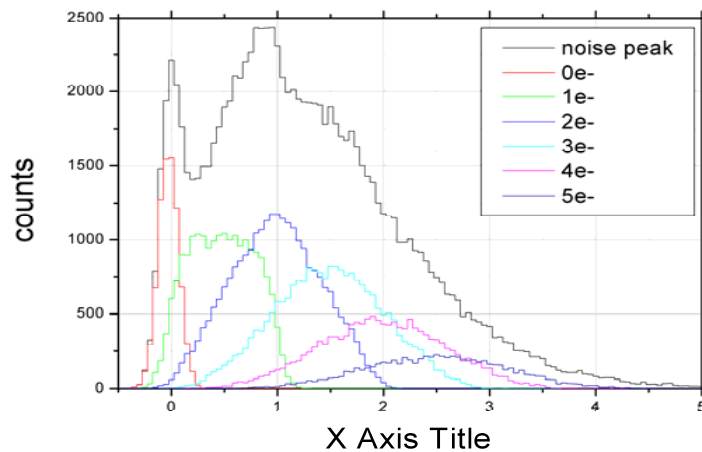
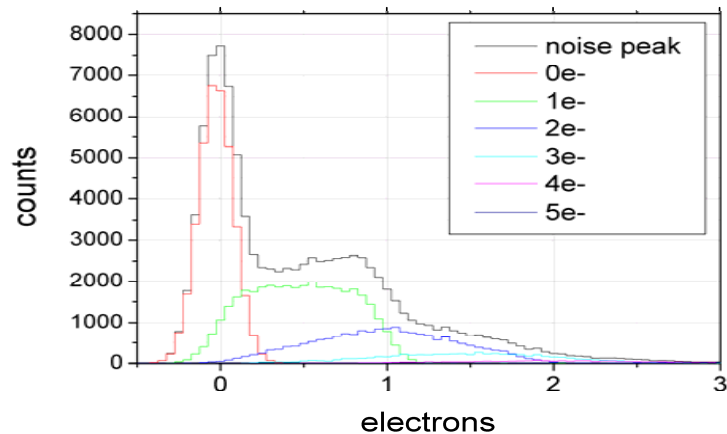
- **White voltage noise:**  
A1=2  
 $a=1.5 \times 10^{-16} \text{ V}^2/\text{sqrt}(\text{Hz})$
- **1/f noise**  
A2=1.26  
 $a_f=4.5 \times 10^{-12} \text{ V}^2$
- **One readout:**  
12.5  $\mu\text{s}$   
White: 1.8 el  
1/f: 1.37 el
- **80 readouts:**  
total time=1.000  $\mu\text{s}$   
White: 0.18 el. r.m.s.  
1/f: 0.16 el  
Total noise: 0.2 el. r.m.s.
- Readout Speed: 1 KHz

***DePMOS with lower 1/f noise are available***

***They have been implemented in the RNDR geometry and will be tested soon***

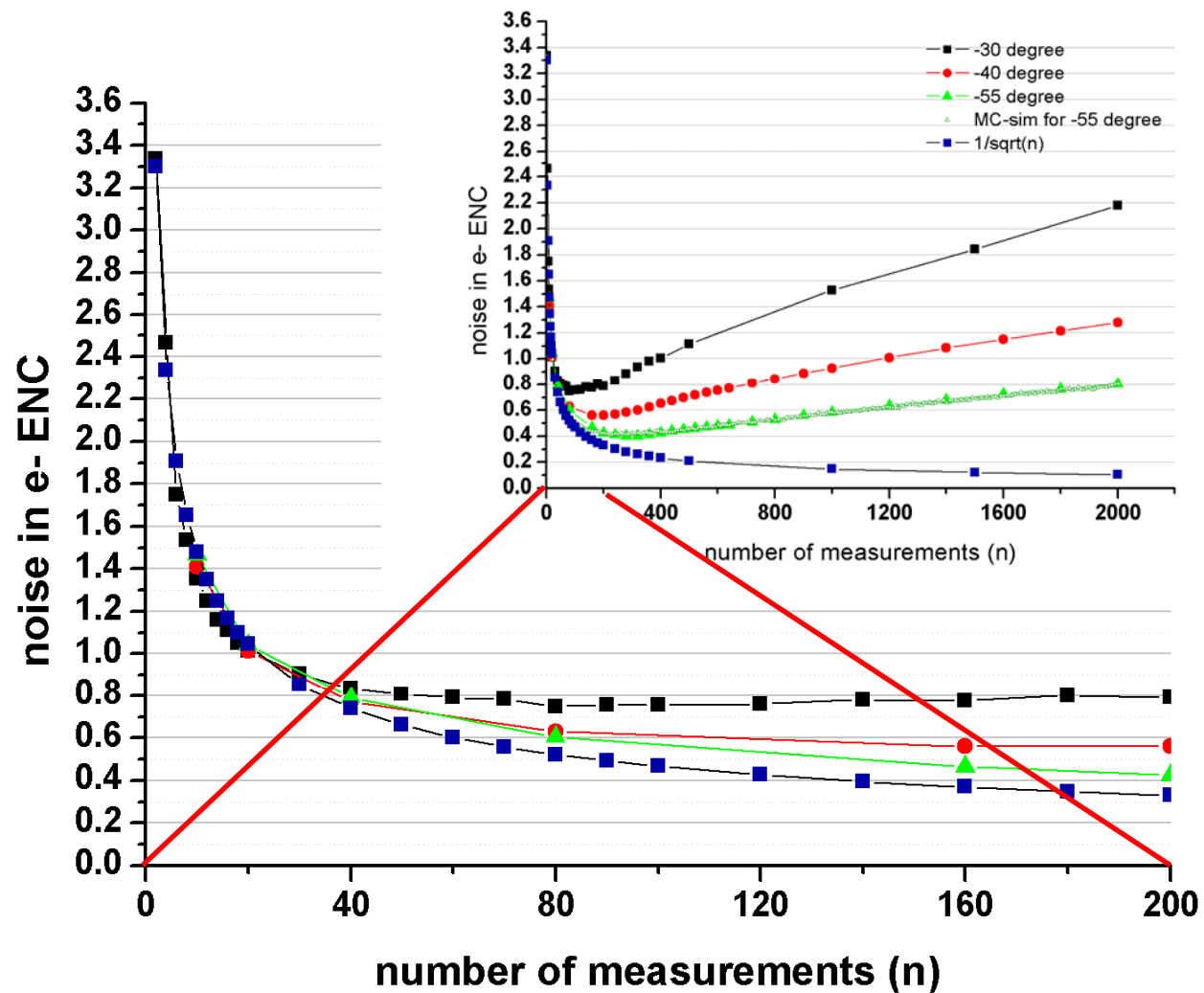
# RNDR-Detectors

## Influence of leakage current

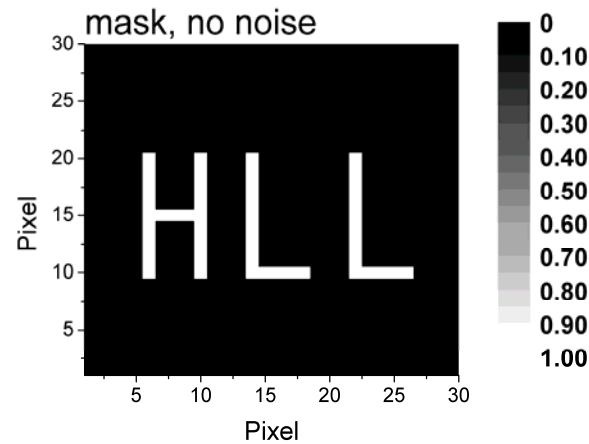


- 100.000 measurement cycles
- 100 loops each
- 3.3 e- noise for one readout cycle (loop)
- In mean **1 electron** during one readout cycle  $t_{acq}$
- In mean **2.5 e-** during  $t_{acq}$

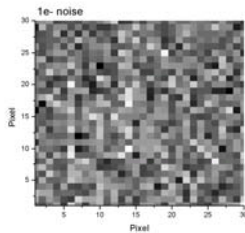
# Noise measurements with HL-devices



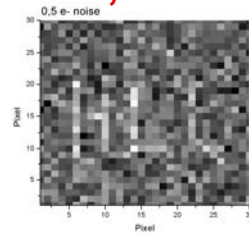
# What does a certain resolution mean in terms of contrast?



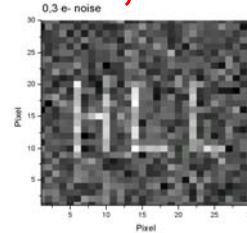
$\sigma = 1 e^-$



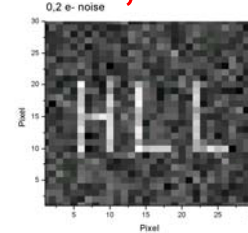
0,5 e-



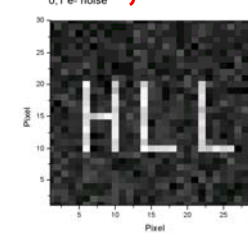
0,3 e-



0,25e-

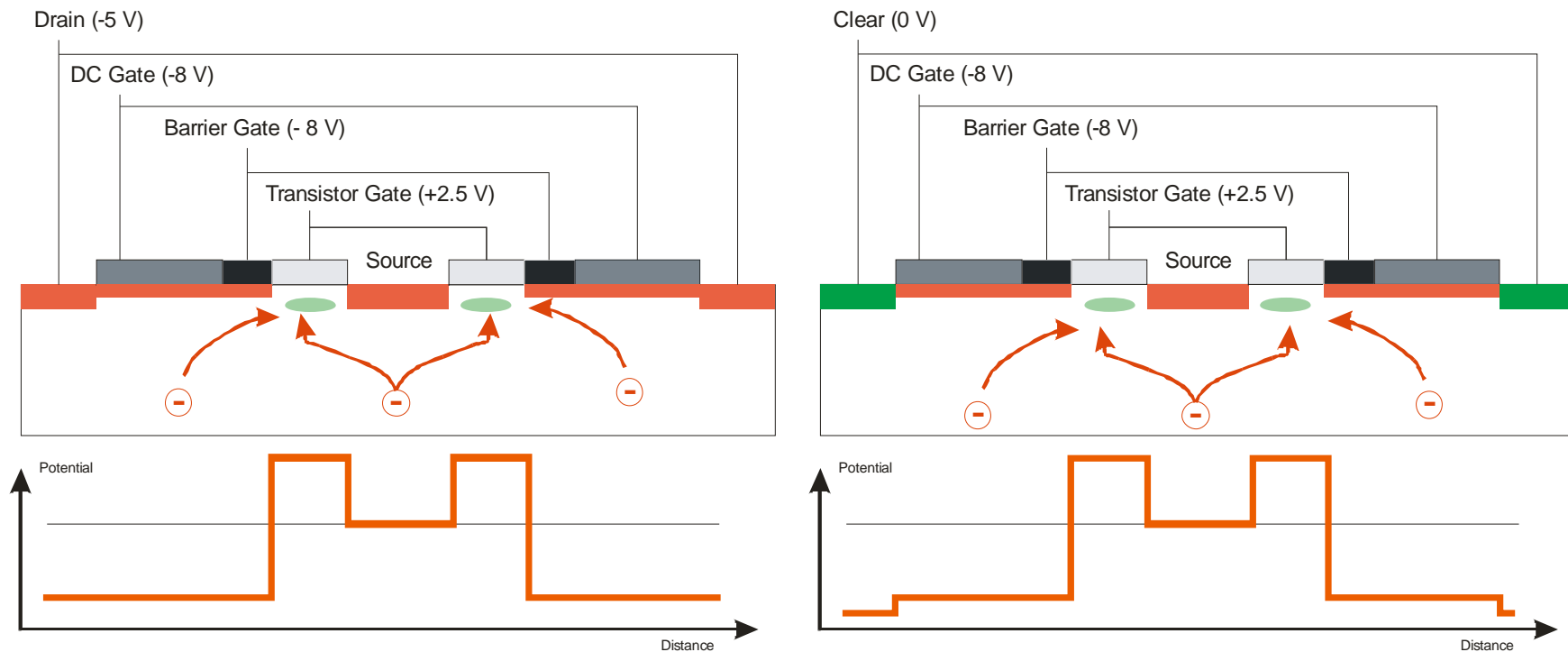


0,1 e-



# Working principle I

## Integration mode

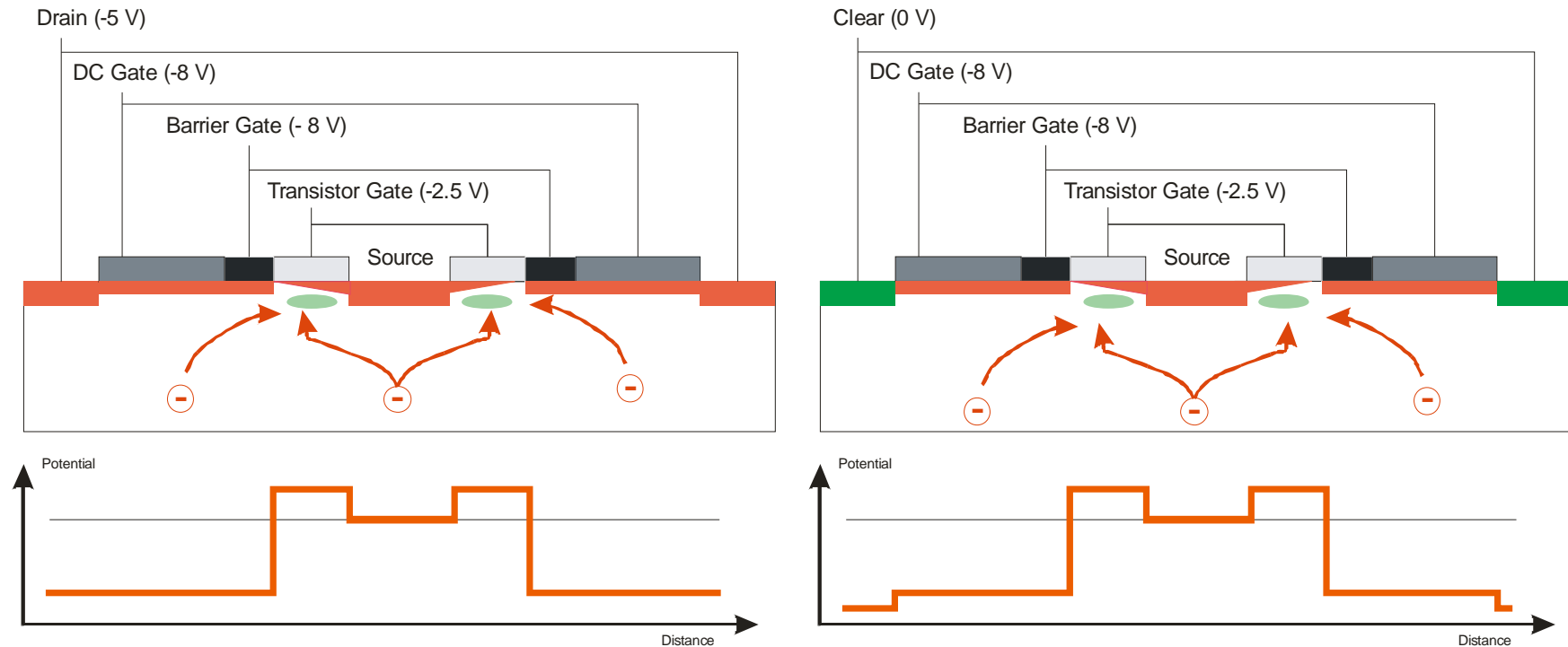


- ❑ „Normal“ DEPFET operation
- ❑ Charge is collected in internal gate
- ❑ No transistor current flowing

# Working principle II



## Readout mode



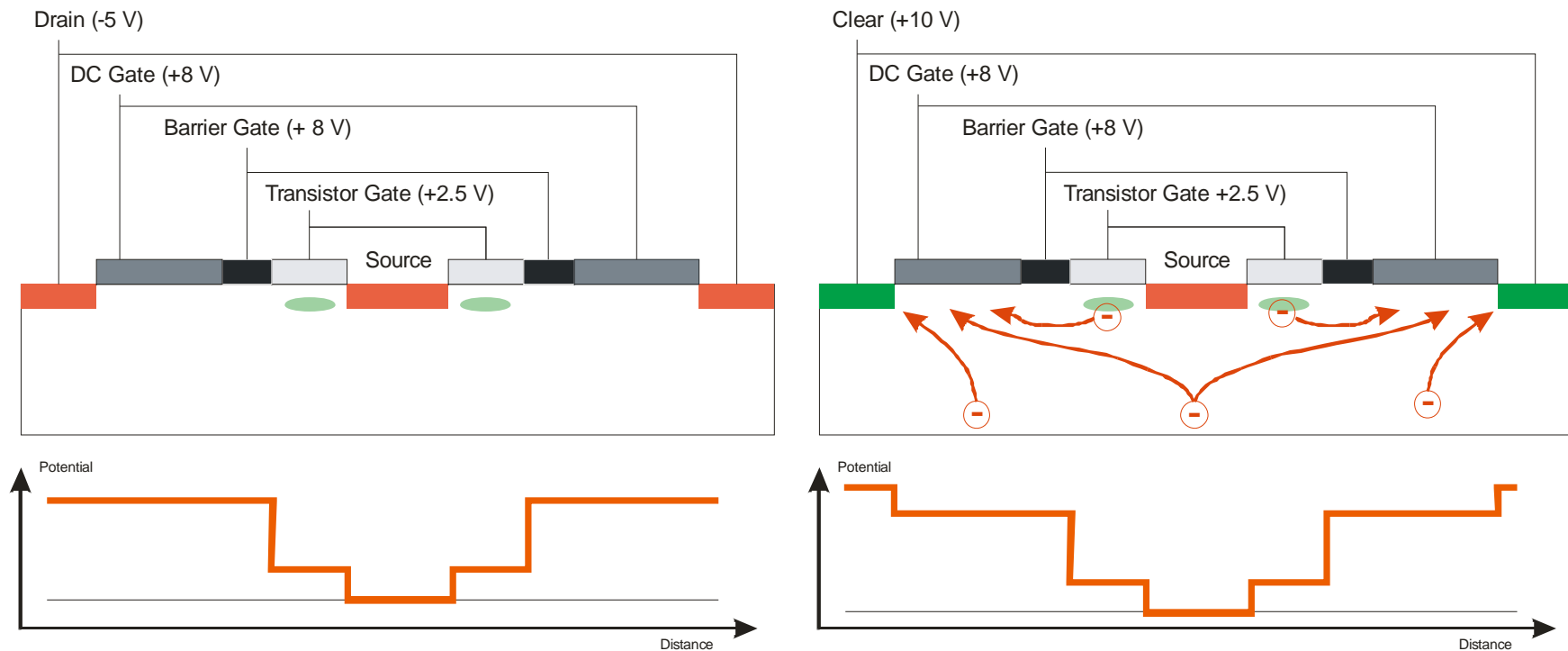
- ❑ „Normal“ DEPFET operation
- ❑ DCGate generates „dynamic drain“



# Working principle III



## Clear mode

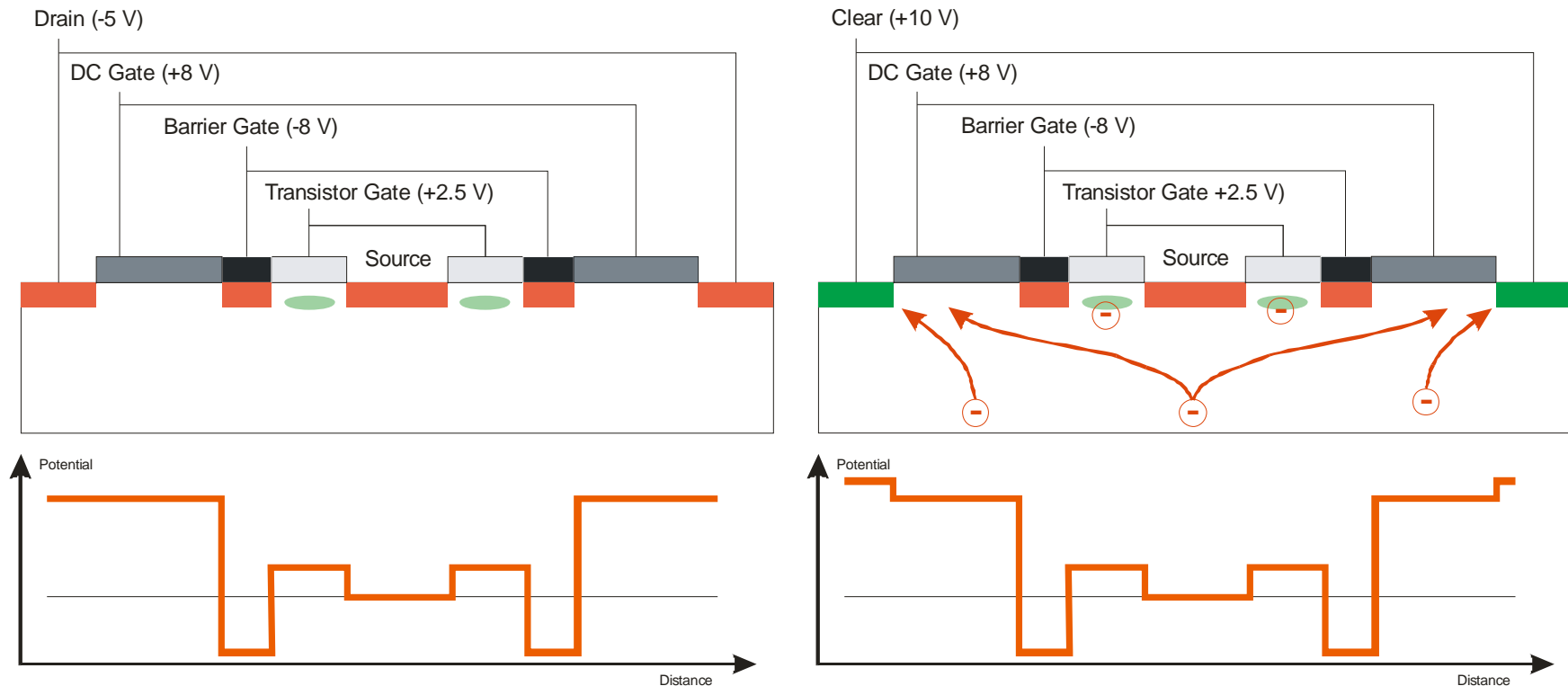


- ❑ Normal DEPFET clear
- ❑ All charge is removed from internal gate
- ❑ Charge from bulk directly drifts to clear

# Working principle IV

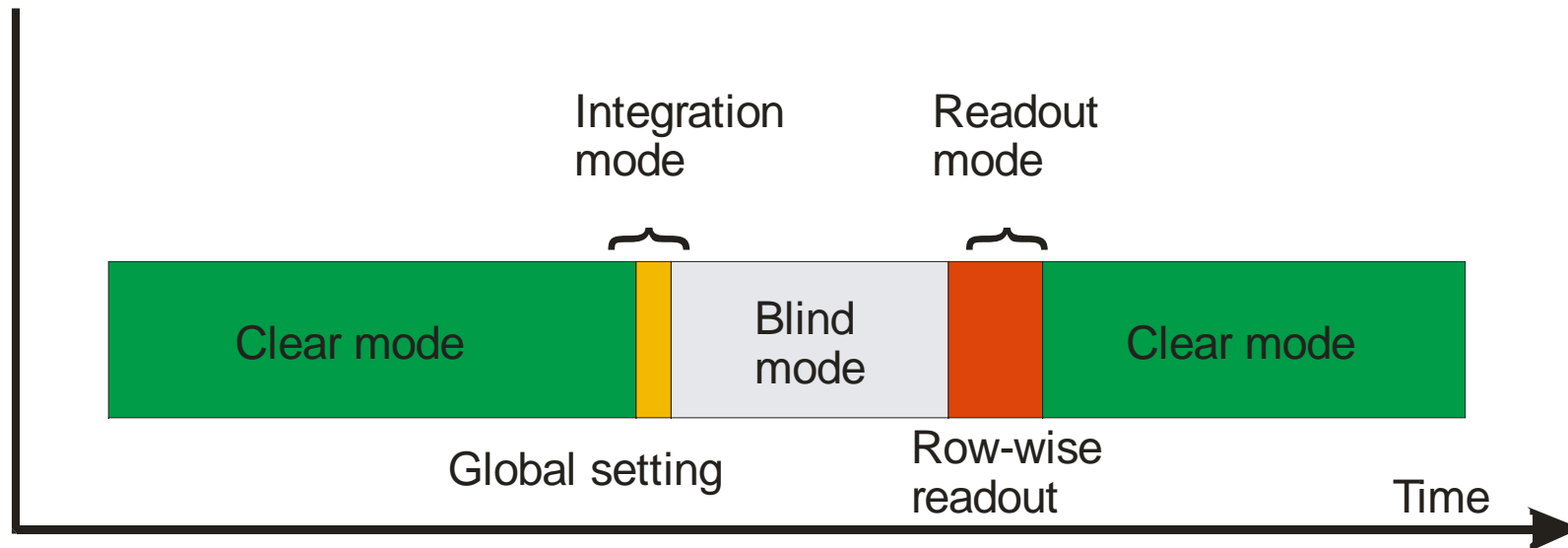


## Blind mode



- New operating state
- Charge from the bulk drifts to the clear contact
- Barrier keeps charge in the internal gate

# Operation cycle



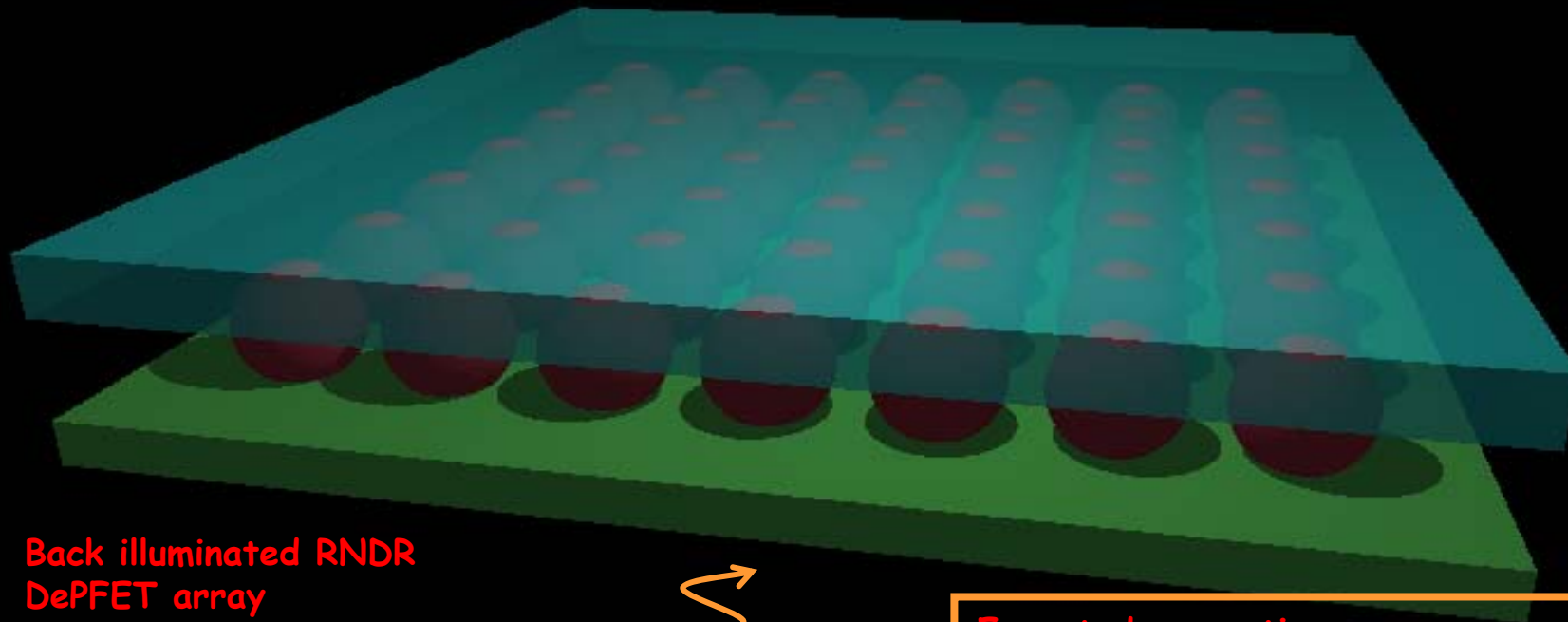
## Applications:

- Selective tagging of time-discrete signal
- Scanning of time continuous, periodic signals
- Run-time detection
- Fluorescence light detection
- etc...

**Expected properties:**

- ❑ 130 nm process
- ❑ bonding area:  $30 \times 30 \mu\text{m}^2$
- ❑ pixel size  $50 \times 50 \mu\text{m}^2$
- ❑ amplifier area  $40 \times 40 \mu\text{m}^2$

**High Speed CMOS amplifier array**



**Back illuminated RNDR  
DePFET array**

incident photon

**Expected properties:**

- ❑ pixel size:  $48 \mu\text{m}$ , format:  $256 \times 256$
- ❑ signal processing:  $10 \mu\text{s}$  per loop
- ❑ frame rate: 1.200 fps
- ❑ NIR sensitive
- ❑ Readout noise:  $0.25 e^-$  @ 64 loops at  $-40^\circ \text{C}$

# Conclusions

- We have presented a concept for a system with single optical photon resolution capability based on a linear RNDR-DePMOS amplifier array
- The working principle of the system has been experimentally demonstrated using:
  - A prototype of a single RNDR-DEPMOS device
  - A prototype of a Multi-channel Low-noise ASIC performing a Trapezoidal Filtering
- A readout noise of 0.18 el. r.m.s. has been obtained at  $-40\text{ }^{\circ}\text{C}$  showing single photon resolution and confirming theoretical predictions
- A fast gating was implemented and experimentally verified
- Using a new kind of RNDR DePMOS device already partially characterized a resolution of 0.25 el. r.m.s. is foreseen with a readout speed of 1.2 kHz
- An optimized circuit already implemented will allow to readout the two DePMOSs of the RNDR device simultaneously, reducing the total readout time of a factor 2